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## Multimedia Video Coding & Architectures (5LSE0), Module 03

### Scalar Quantization

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slides version 1.0

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## Multimedia Video Coding and Architectures (5LSE0), Module 03

### Mod 03, Preliminaries

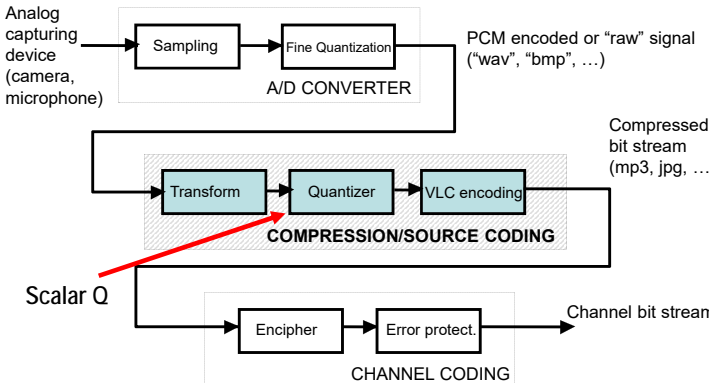
#### Brief introduction: where are we...?

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## System Overview



The diagram illustrates the system overview. It starts with an 'Analog capturing device (camera, microphone)' which feeds into a box labeled 'A/D CONVERTER'. Inside this box are 'Sampling' and 'Fine Quantization' blocks. The output is a 'PCM encoded or "raw" signal ("wav", "bmp", ...)'. This signal then goes into a box labeled 'COMPRESSION/SOURCE CODING', which contains 'Transform', 'Quantizer', and 'VLC encoding' blocks. A red arrow labeled 'Scalar Q' points to the 'Quantizer' block. The output is a 'Compressed bit stream (mp3, jpg, ...)'. This bit stream then goes into a box labeled 'CHANNEL CODING', which contains 'Encipher' and 'Error protect.' blocks. The final output is a 'Channel bit stream'.

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## Why can Signals be Compressed? – (1)

*Because ~~signal amplitudes~~ are statistically redundant*  
(When Quantizing:) *quantizer representation levels*

Question 1:  
What is the *shortest* average codeword length that one can achieve for a given signal (or "source")?


$$H(S) = -\sum_{i=1}^N P_s(s_i) \log_2 [P_s(s_i)] \quad (\text{bit})$$

$$H(X) = -\int_{-\infty}^{\infty} p_X(x) \log_2 [p_X(x)] dx$$

Question 2:  
How did you obtain those codewords?

*Huffman Coding*  
*Run-length coding*

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## Why can Signals be Compressed? – (2) <sup>5</sup>

Because signal amplitudes are mutually dependent

Question 1:

What is the *best possible* exploitation of the correlation (dependencies) in natural signals?  
(Rate-Distortion Theory)

Question 2:

How do we *implement* a system that exploits the correlation in natural signals?

(Compression algorithms: - DPCM  
- Subband/wavelet  
- Transform/DCT  
- Motion compensation)

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## Why can Signals be Compressed? – (3) <sup>6</sup>

Because infinite accuracy of signal amplitudes is (perceptually) irrelevant

Question 1:

What is the *best possible* trade-off between required bit rate and resulting distortion?

(Rate-Distortion Theory)

Question 2:

How do we *implement* a system that gives us that best possible trade-off?

(Scalar and Vector Quantization Theory)

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5LSE0 - Mod 03

Part 1

Characterization of Quantizers

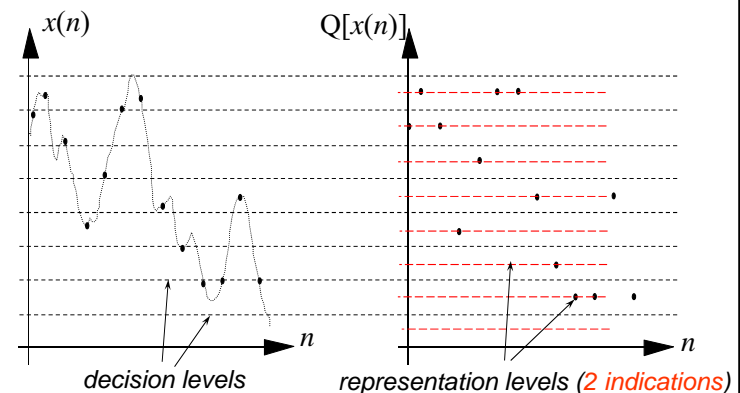
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## Quantization Process – (1) <sup>8</sup>



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## Quantization Process – (2)

- \* **Quantization:** Map a continuous-valued signal value  $x(n)$  onto a (limited set of) discrete-valued signals values  $y(n)$ :
 
$$y(n) = Q[x(n)]$$
 such that  $y(n)$  is good approximation of  $x(n)$ .
- \* **Important:** # bits to represent  $y(n)$ .  
(= average codeword length per representation level)
- \* **Design/Optimization Problem**
  - What positions of the "decision levels"
  - What positions of the "representation levels"

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## Quantization Process – (3)

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## Formal Quantization Definition

- \* Given amplitude  $x$ , divide domain of  $x$  into  $K$  non-overlapping intervals  $S_k$ :
 
$$S_k = \{x \mid x_k < x \leq x_{k+1}\} \quad k = 1, 2, \dots, K$$
- \* If  $x$  falls in  $S_j$  then it is represented by  $y_j$ .
- \* Quantization gives *coding errors or quantization noise*

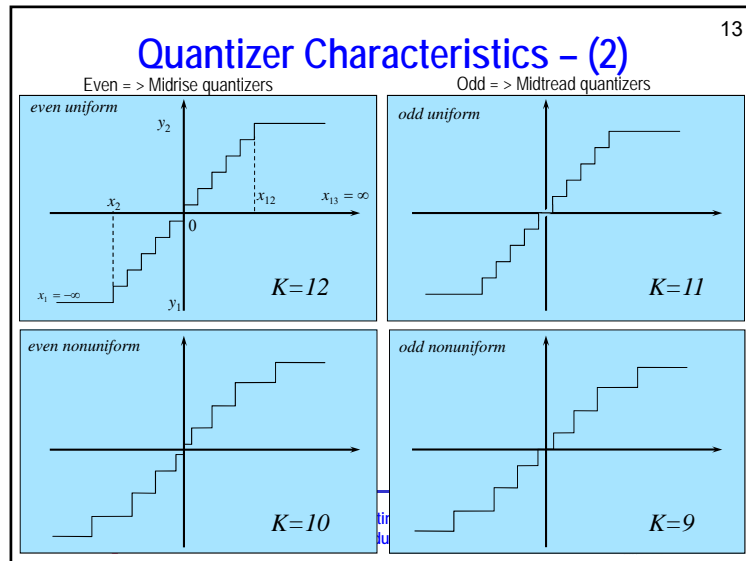
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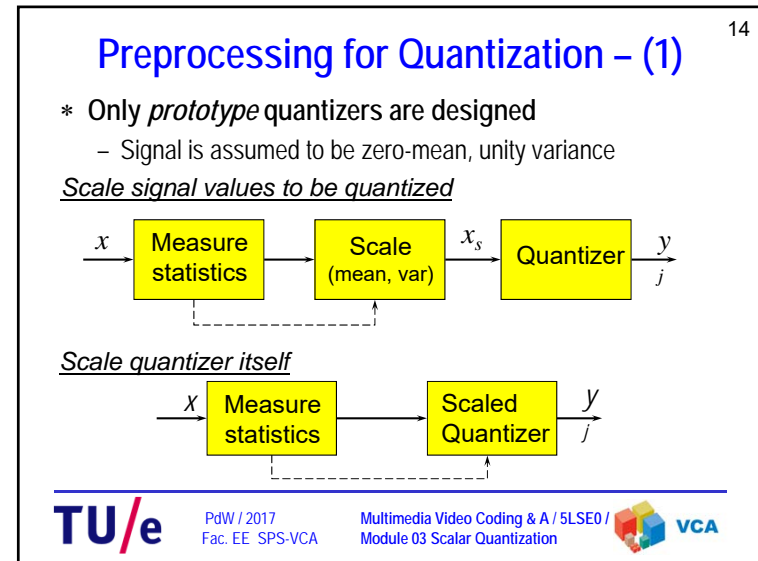
## Quantizer Characteristics – (1)

- \* Signals values are assumed to lie symmetrically around zero
  - ⇒ Quantizers are usually *symmetric about origin*
- \* Quantizer choices:
  - Odd or even number of levels? } "structure"
  - Uniform or non-uniform quantizers? }
  - Way of optimizing the quantizer? } "design"

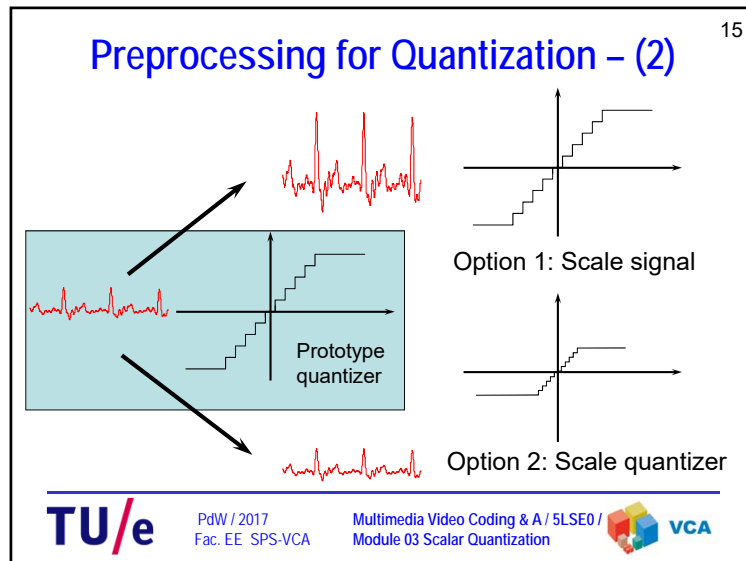
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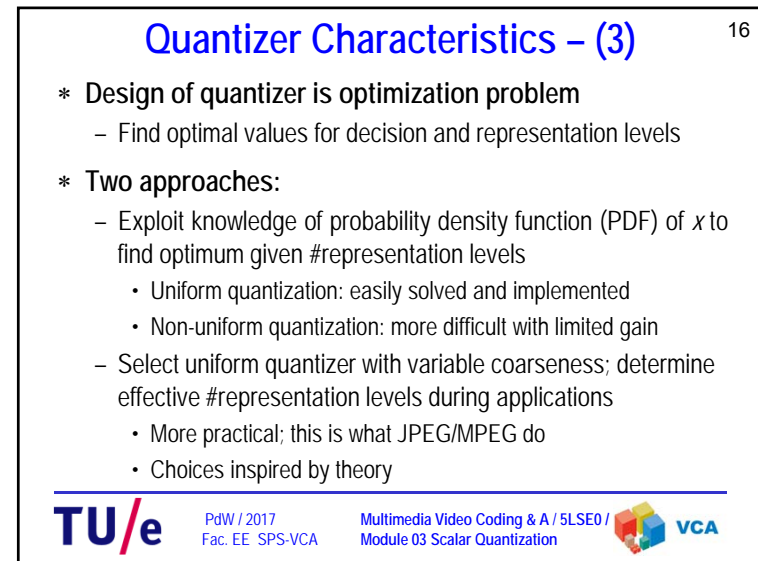
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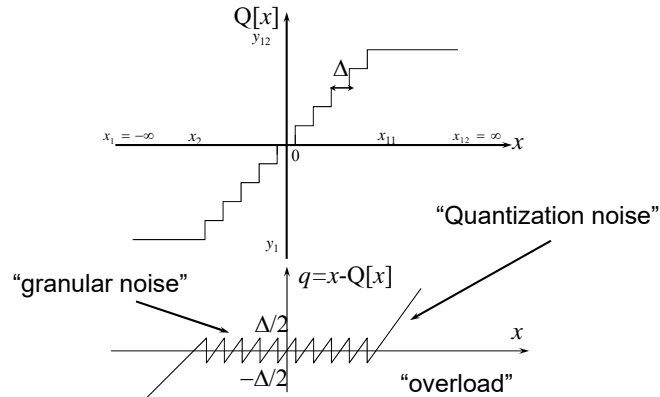
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## Quantization Noise / types – (1)

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## Quantization Noise / Variance – (2)

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- \* Quantification of (average) quantization error:  
*Variance of the quantization noise:*  $q = x - Q[x]$
- \* Need to model probability density of  $x$ :  $p_X(x)$
- \* Quantization noise variance:

$$\sigma_q^2 = \int_{-\infty}^{\infty} \underbrace{(x - Q[x])^2}_{\text{Amount of error}} \underbrace{p_X(x)}_{\text{Probability of amount}} dx$$

Amount of error    Probability of amount

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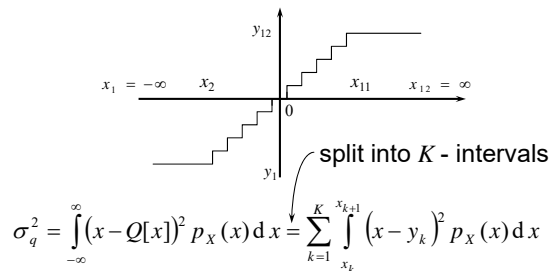
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## Quantization Noise / Variance – (3)

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$$\sigma_q^2 = \int_{-\infty}^{\infty} (x - Q[x])^2 p_X(x) dx = \sum_{k=1}^K \int_{x_k}^{x_{k+1}} (x - y_k)^2 p_X(x) dx$$

Normalized measure: Signal-to-Noise-Ratio (SNR):

$$SNR = 10 \log_{10} \left( \frac{\sigma_x^2}{\sigma_q^2} \right) \text{ (dB)}$$

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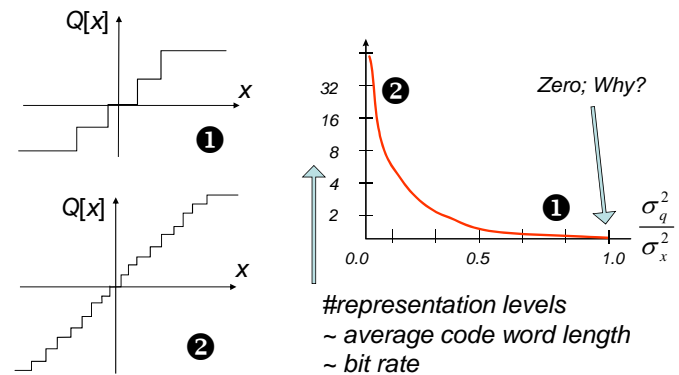
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## Bit Rate versus Distortion – (1)

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#representation levels  
~ average code word length  
~ bit rate

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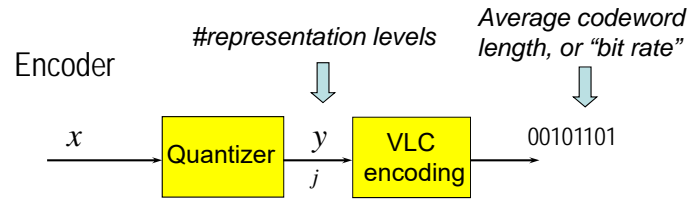
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## Terminology for using quantizers

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- \* Representation levels coded in  $\lceil \log_2(K) \rceil$  bits:  
*Fixed-rate quantization*
- \* Representation levels coded in  $H(Y)$  bits:  
*Quantization with entropy encoding*
- \* (Later: *Entropy-constrained quantization*)

## Bit Rate versus Distortion – (2)

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- \* Performance of quantizer is determined by
  - the number of representation levels (bit rate or average codeword length  $R$ )
  - the quality  $\sigma_q^2$  or SNR

- \* Fixed-rate quantizer design:

For given  $K$ , find the quantizer characteristic with smallest  $\sigma_q^2$

## 5 LSE0 - Mod 03, Part 2 Uniform Quantization

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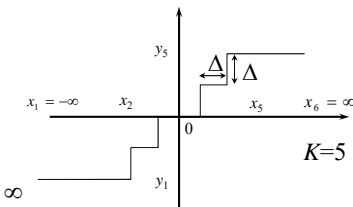
## R-D Optimal Design: Uniform Quantizer – (1)

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- \* Given:

$$x_1 = -\infty$$

$$x_{L+1} = \infty$$



$$x_{k+1} - x_k = \Delta$$

$$y_k = \frac{(x_{k+1} + x_k)}{2}$$

- \* Find  $\Delta$  such that  $\sigma_q^2$  is minimized

## R-D Optimal Design: Uniform Quantizer – (2)

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$$\min_{\Delta} \sum_{k=1}^K \int_{x_k}^{x_{k+1}} (x - y_k)^2 p_X(x) dx$$

For values of  $K > 3$ , this requires numerical optimization

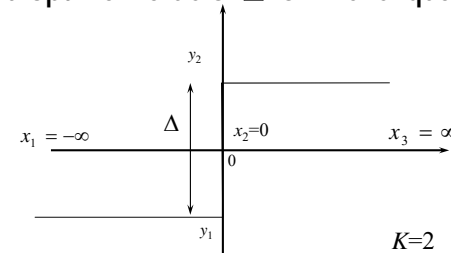
R bit/sample)	$\Delta/\sigma_x = 1$			SNR (dB)		
	Uniform	Gaussian	Laplace	Uniform	Gaussian	Laplace
1	1.732	1.596	1.414	6.02	4.40	3.01
2	0.866	0.996	1.087	12.04	9.25	7.07
3	0.433	0.586	0.731	18.06	14.27	11.44
4	0.217	0.335	0.461	24.08	19.38	15.96
5	0.108	0.188	0.280	30.10	24.57	20.60
6	0.054	0.104	0.166	36.12	29.83	25.36
7	0.027	0.057	0.096	42.14	35.13	30.23
8	0.013	0.031	0.055	48.17	40.34	35.14

smaller  $D$ : "finer" quantizer more difficult to quantize

## Example: Two-Level Quantizer – (1)

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\* Find optimal value of  $\Delta$  for 2 level quantizer



$$\sigma_q^2 = \sum_{k=1}^K \int_{x_k}^{x_{k+1}} (x - y_k)^2 p_X(x) dx = \int_{-\infty}^0 (x + \frac{\Delta}{2})^2 p_X(x) dx + \int_0^{\infty} (x - \frac{\Delta}{2})^2 p_X(x) dx$$

## Two-Level Quantizer – (2)

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$$\sigma_q^2 = \int_{-\infty}^0 (x + \frac{\Delta}{2})^2 p_X(x) dx + \int_0^{\infty} (x - \frac{\Delta}{2})^2 p_X(x) dx$$

$$= \sigma_x^2 + \frac{\Delta^2}{4} - 2\Delta \int_0^{\infty} x p_X(x) dx$$

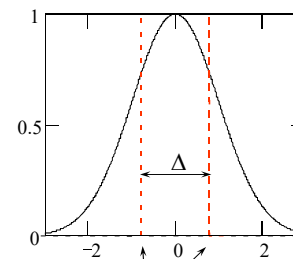
$$\Rightarrow \Delta_{\text{optimal}} = \min_{\Delta} \sigma_q^2 = 4 \int_0^{\infty} x p_X(x) dx = 2E[|X|]$$

PDF	$\Delta/\sigma_x$	$\sigma_q^2$	SNR (dB)
Uniform	1.732	0.250	6.02
Gaussian	1.596	0.363	4.40
Laplace	1.414	0.500	3.01
Gamma	1.154	0.667	1.76

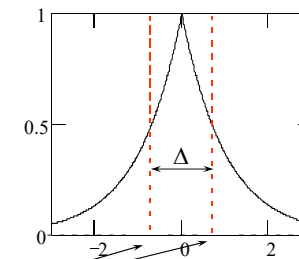
## Two-Level Quantizer – (3)

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a. Gaussian PDF



b. Laplace PDF



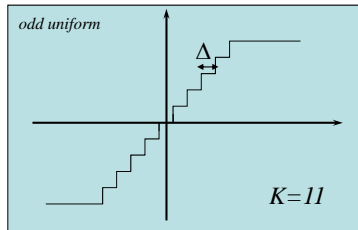
quantization levels

## Implementation of Uniform Quantizer

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### \* The quantizer

- Odd
- Uniform
- $K = 11$  outp.levels



can be implemented by:

$$Q[x] = y_j = \Delta \operatorname{nint}\left(\frac{x}{\Delta}\right) = \Delta \left\lfloor \frac{x}{\Delta} + \frac{1}{2} \right\rfloor$$

$$j = \operatorname{nint}\left(\frac{x}{\Delta}\right) = \left\lfloor \frac{x}{\Delta} + \frac{1}{2} \right\rfloor$$

## 5LSE0 Mod 03, Part 3 Non-Uniform Quantization

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## Non-Uniform Quantizer / 2 Approaches

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### 1. Preprocessing of $x$ by non-linear function followed by uniform quantizer:

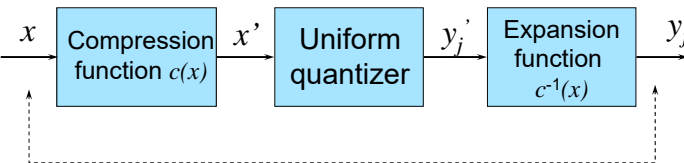
- Companding (compression-expanding)
- Simple implementation
- Popular for audio: logarithmic curves
  - A-law (Europe)
  - and  $\mu$ -law (USA, Japan)

### 2. Lloyd-Max quantizers, minimization of $\sigma_q^2$

- Complex design
- More complex implementation than uniform quantizer
- Additional gain?

## Companding – (1)

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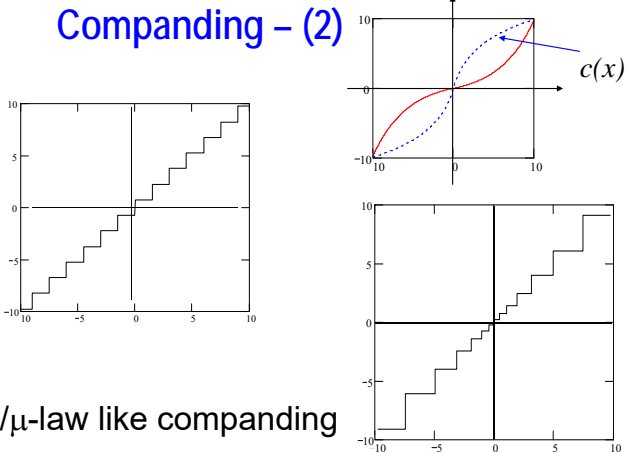


Non-uniform behavior

Uniform quantizer is embedded between  
two non-uniform functions




## Compinging - (2)



A/ $\mu$ -law like companding

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## A- and $\mu$ -Law Definition

**A-Law**


$$c(x) = \begin{cases} x_{\max} \frac{ax / x_{\max}}{1 + \ln(a)} & 0 \leq x/x_{\max} \leq 1/a \\ x_{\max} \frac{\ln(a|x/x_{\max}|)}{1 + \ln(a)} \text{sgn}(x) & 1/a < x/x_{\max} \leq 1 \end{cases}$$

**$\mu$ -Law**

$$c(x) = x_{\max} \frac{\ln(1 + \mu|x/x_{\max}|)}{\ln(1 + \mu)} \text{sgn}(x)$$

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## Lloyd-Max Quantizer - (1)

\* Minimizes quantization noise variance, without enforcing any structure onto decision thresholds and representation levels

$$\min \sigma_q^2 = \min \sum_{k=1}^K \int_{x_k}^{x_{k+1}} (x - y_k)^2 p_X(x) dx$$


for:

$$x_k \quad k = 2, 3, \dots, K$$

$$y_k \quad k = 1, 2, \dots, K$$

\* except for symmetry of the quantizer

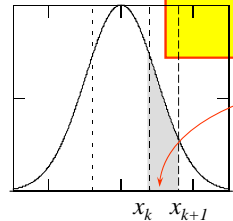
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## Lloyd-Max Quantizer - (2)


\* General solution is given by implicit expressions:

$$x_k = \frac{1}{2}(y_{k-1} + y_k) \quad y_k = \frac{\int_{x_k}^{x_{k+1}} xp_X(x) dx}{\int_{x_k}^{x_{k+1}} p_X(x) dx}$$



- Note the structure of the formulas
  - Middle of two representation levels
  - Weighted average decision levels

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## Lloyd-Max Quantizer – (3)

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### \* Property of solution:

- if  $x_k$  are known  $\Rightarrow y_k$  are known
- if  $y_k$  are known  $\Rightarrow x_k$  are known

### \* Iterative design necessary

initial choice, "seed"

$$x_1 = -\infty$$

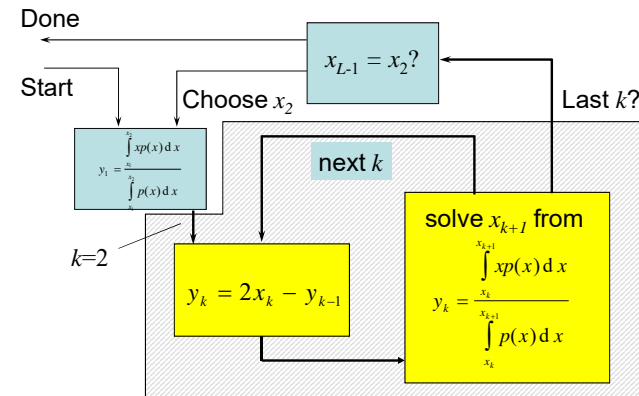
Assume:  $x_2 = g \Rightarrow y_1 = \frac{\int_{x_1}^{x_2} xp(x)dx}{\int_{x_1}^{x_2} p(x)dx} \Rightarrow y_2 = 2x_2 - y_1$

Solve  $x_3$  from:  $y_2 = \frac{\int_{x_2}^{x_3} xp(x)dx}{\int_{x_2}^{x_3} p(x)dx} \Rightarrow y_3 = 2x_3 - y_2$

etcetera

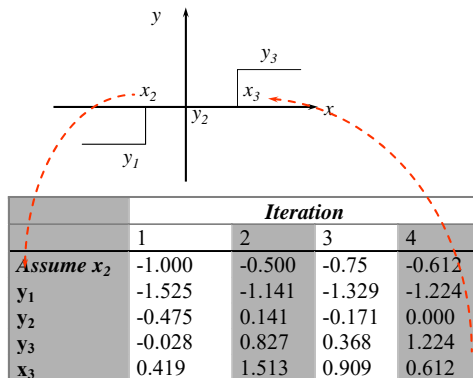
## Lloyd-Max / Iterative Solution Scheme

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## Lloyd-Max / Example iterative Q design

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## Examples of Lloyd-Max Quantizers

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• [Example from VcDemo Quantizer](#)

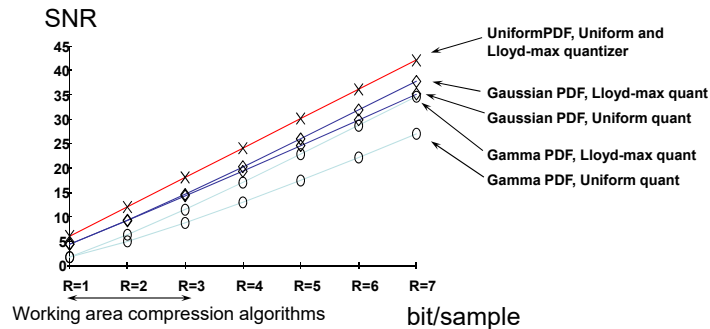
\* For *uniform PDF*, Lloyd-Max quantizer is *uniform*

PDF	$k$	$K=2, 1 \text{ bit/sample}$		$K=4, 2 \text{ bit/sample}$		$K=8, 3 \text{ bit/sample}$		$K=16, 4 \text{ bit/sample}$	
		$x_k$	$y_k$	$x_k$	$y_k$	$x_k$	$y_k$	$x_k$	$y_k$
Gaussian	1	0.000	0.798	0.000	0.453	0.000	0.245	0.000	0.128
	2			0.982	1.510	0.501	0.756	0.258	0.388
	3					1.050	1.344	0.522	0.657
	4					1.748	2.152	0.800	0.942
	5							1.099	1.256
	6							1.437	1.618
	7							1.844	2.069
	8							2.401	2.733
Laplace	1	0.000	0.707	0.000	0.402	0.000	0.233	0.000	0.124
	2			1.127	1.834	0.533	0.833	0.264	0.405
	3					1.253	1.673	0.567	0.729
	4					2.380	3.087	0.920	1.111
	5							1.345	1.578
	6							1.878	2.178
	7							2.597	3.017
	8							3.725	4.432

## Uniform quantizer versus Lloyd-Max

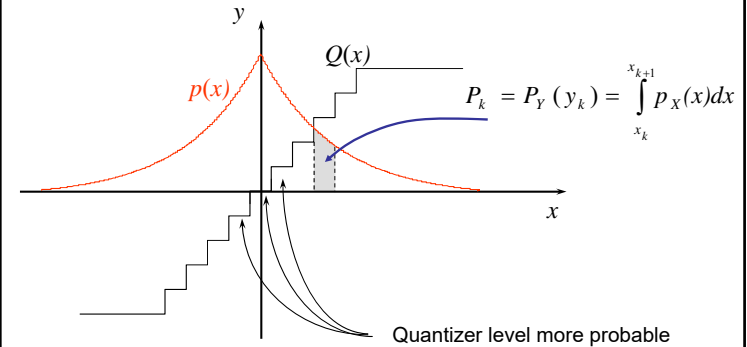
41

For relevant bit rates, Lloyd-Max does often not pay off.



## Quantization with Entropy Coding – (1)

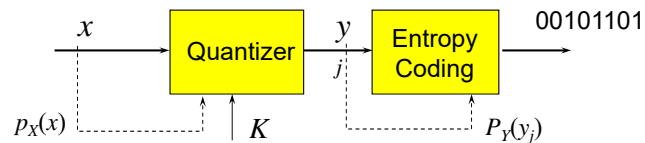
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• [Example from VcDemo Quantizer](#)

## Quantization with Entropy Coding – (2)

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1. Quantizer is optimized for *given* #representation levels  $K$
2. Entropy coder (VLC) is optimized for given probabilities of  $y_j$ , yielding average codeword length  $L$

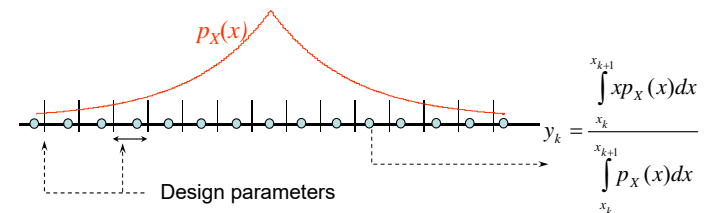
But, the combination is *not* optimal

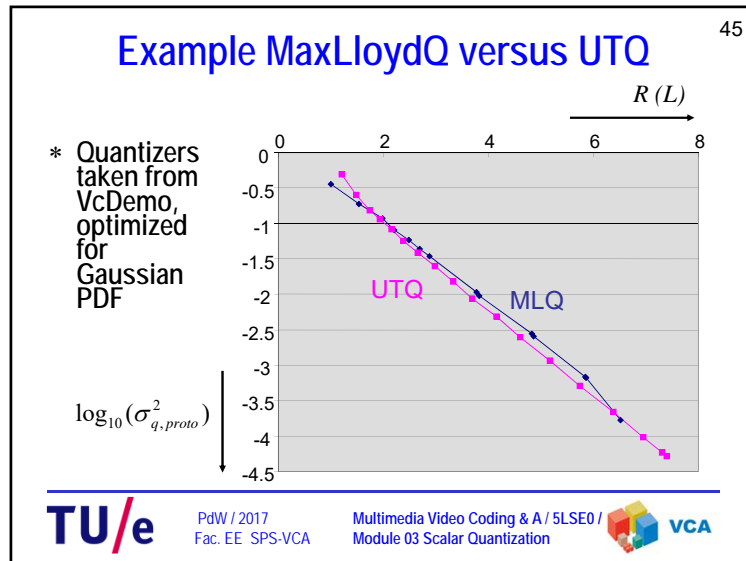
- other quantizer / VLC combinations exist that gives a smaller quantization error variance for same average codeword length

## Entropy-Constrained Quantization

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- \* To find the overall optimal result, quantizer and entropy coder must be *jointly* designed
  - Complex optimization problem
  - Reasonable approximations are obtained by *Uniform Threshold Quantizers (UTQ)*





## 5LSE0 – Mod 03, Part 4

### Quantization in practical cases

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### Quantization in Practice – (1)

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- \* **Uniform quantizers are preferred**
  - Implementation and limited loss compared to Lloyd-Max
  - Easily scalable (one parameter: step size  $\Delta$ )
- \* **Odd quantizers are often preferred over even** because of the presence of a representation level at zero
  - In good compression scheme many (near-)zero values occur
  - Zeroes efficiently coded by an entropy coder (run length coding)
- \* **Audio:** Companding is usual
- \* **Image/video coding :** No companding
  - Uniform quantizer with *deadzone* is typical

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### Quantization in Practice – (2)

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$Q[x]$

$x$

deadzone

$2\Delta$

VLC

$\Delta$	1000
$2\Delta$	1110
$3\Delta$	110
$4\Delta$	0
$5\Delta$	101
$6\Delta$	1111
$7\Delta$	10011

**Deadzone:**

- Improves noise robustness of coding system
- “Stimulates” truncation to zero: can be coded efficiently

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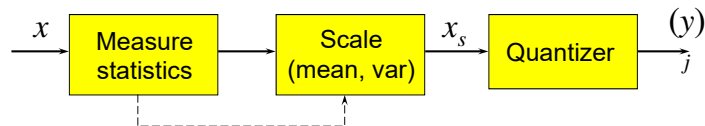
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## Overall Quantization Error – (1)

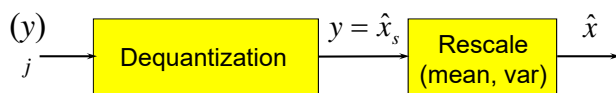
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- \* Only *prototype* quantizers are designed with  $\sigma_q^2$

Sending side



Receiving side



## Overall Quantization Error – (2)

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- \* Scaling procedure:  $x_s = \frac{x - \mu_x}{\sigma_x}$
- \* Rescaling procedure:  $\hat{x} = \sigma_x y + \mu_x = \sigma_x \hat{x}_s + \mu_x$
- \* Overall effect: Quantization noise variance is scaled with:  $\sigma_x^2$

$$\sigma_q^2 = E[(\hat{x} - x)^2] = E[(\sigma_x \hat{x}_s - \sigma_x x_s)^2] = \sigma_x^2 E[(\hat{x}_s - x_s)^2] = \sigma_x^2 \sigma_{q, \text{prototype}}^2$$

⇒ *Quantization noise variance is linearly proportional to the variance of the signal to be quantized*

The following sheets were used in previous classes and are included for reference only

## Quantizer Example - (2)

[Back to ML Quantizers](#) [Back to Prob. Calculation](#)

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	Reprs. Level	Probability	Codewords (positive)	Codewords (negative)
1	±9.247	0.00007	00010101101010	00010101101011
2	±6.875	0.00031	0001010110111	0001010110100
3	±5.519	0.00077	00010101111	00010101100
4	±4.562	0.00152	0001010000	0001010001
5	±3.823	0.00269	000101001	000101010
6	±3.217	0.00444	00011010	00011011
7	±2.703	0.00699	1011010	1011011
8	±2.253	0.01068	0001011	0001100
9	±1.855	0.01590	000111	101100
10	±1.497	0.02318	10111	000100
11	±1.175	0.03316	01000	01001
12	±0.884	0.04658	1100	1101
13	±0.622	0.06441	0101	1010
14	±0.388	0.08797	111	0000
15	±0.180	0.11987	011	100
16	0.000	0.16292	001	

- \* Simple binary coding would require  $\lceil \log_2(31) \rceil = 5$  bits
- \*  $H(S) = 3.876$  bit/repr.level
- \*  $L = 3.912$  bit/repr.level

## Application to Signal Coding

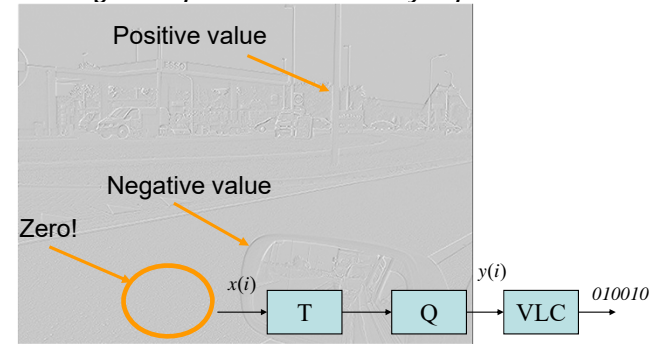
- \* We will see that efficient “transforms” used in compression produce
  - A lot of “zero” values
  - Some (significant) non-zero values
- \* Typical symbol sequences to be coded:
 

“5 1 0 0 0 0 0 0 0 3 0 0 6 0 0 0 0 1 0 0 0 0 ...”

  - Will be done by {zero-run, non-zero symbol} pairs
  - Here: “{0,5}, {0,1}, {7,3}, {2,6}, {4,1}, ...”
  - The pairs will now be assigned a Huffman code

## Why can Signals be Compressed?

Because signal amplitudes are mutually dependent



The following sheets are included for further reference

## Note on Quantization Noise Variance

- \* We wish to calculate the variance of  $q$ :

$$\sigma_q^2 = E[q^2] - \underbrace{E[q]^2}_{=0} = \int_{-\infty}^{\infty} q^2 p_q(q) dq$$

- \* Try to calculate  $p_q(q)$ , but with non-linear relation of  $q=x-Q[x]$ , this is not easy
- \* Much easier, take following short-cut (E-operator):

$$\begin{aligned} \sigma_q^2 &= E[q^2] = E[(x - Q[x])^2] \\ &= \int_{-\infty}^{\infty} (x - Q[x])^2 p_x(x) dx \end{aligned}$$

## Modeling of Quantizers

- \* Quantizer is often the only non-linear element in a compression system
- \* Linear (approximating) models are sometimes used to analyze the performance of a quantizer / compression system

